### **Mobile Group Seminar**

Reduction of AOA estimation error due to perturbation in array response by spatial smoothing preprocessing

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# **Background (1)**

- Subspace-based AOA estimation, e.g. MUSIC/ ESPRIT, requires the precise array response.
- In practice, it is difficult to obtain the precise array response even if the antenna calibration is applied.
- Therefore, methods to compensate the calibration error are required.

# **Background (2)**

How to compensate the error in the array response?

- In 1993, Hari and Gummadavelli proposed the performance analysis of subspace methods for estimating AOA due to SSP in the presence of array model errors.
  - SSP can improve the performance of ESPRIT and Min-Norm in the presence of errors in the array response, but it is not so for MUSIC.
  - The theoretical expressions were well derived, but complicated and verified by only simulation.
- Objective

To investigate the possibility of SSP to reduce the random error in the array response in real systems.

# **Experiment (1): Anechoic chamber**



curvature of spherical wavefront is compensated to far-field planar wavefront before data processing.

### **Experiment (2): Specification of experiment**



#### Antenna array

Shape of array	Uniform linear array
Number of elements	10
Element spacing	0.8 λ
Antenna element	Patch antenna
Antenna gain	7 dBi

#### Transmitter

Frequency	1.74 GHz
Antenna	Horn antenna
Antenna gain	14.67 dBi
Tx power	-30 dBm
Modulation	GMSK



### Experiment (3): Open site



Moving Tx where AOA at Rx are known in range -6 to 6 degrees.

### **Signal Model**

The array output vector can be modeled as

 $\mathbf{x}(t) = \mathbf{a} s(t) + \mathbf{n}(t)$ 

1) $d\sin\theta$  $d\sin\theta$ where s(t) is an arriving signal  $\boldsymbol{n}(t)$  is a noise vector  $a = [1, e^{j\phi_1}, e^{j\phi_2}, ..., e^{j\phi_M}]^T$  $x_1(t) = x_2(t) = x_k(t)$  $x_{M}(t)$ and  $\phi_m$  is the nominal phase of the  $m^{th}$  element;  $\phi_m = (m-1)\omega$ 

s(t)

θ

 $\omega = 2\pi d \sin(\theta) / \lambda$  is the phase difference between adjacent elements. The output covariance matrix of the single source can be written as

$$\boldsymbol{R} = E[\boldsymbol{x}(t)\boldsymbol{x}^{H}(t)] = \sigma_{s}^{2}\boldsymbol{a}\boldsymbol{a}^{H} + \sigma_{n}^{2}\boldsymbol{I}$$
$$= \lambda_{s}\boldsymbol{u}_{s}\boldsymbol{u}_{s}^{H} + \sigma_{n}^{2}\boldsymbol{I}$$

AOA estimation using subspace-based approaches: ESPRIT and MUSIC.

### **Spatial smoothing preprocessing:** Forward-only SSP



The output vector of the  $l^{th}$  subarray,  $\mathbf{x}_{l}^{f}(t) = \mathbf{a}_{l} s(t) + \mathbf{n}_{l}(t), \ l = 1, 2, ..., L$   $\mathbf{M}$  where  $\mathbf{a}_{l} = [a_{l}, a_{l+1}, ..., a_{l+M_{0}-1}]^{T}$ , and  $\mathbf{n}_{l}(t) = [n_{l}(t), n_{l+1}(t), ..., n_{l+M_{0}-1}(t)]^{T}$ ,

The covariance matrix of the  $l^{th}$  subarray is given by

$$\boldsymbol{R}_{l}^{f} = E[\boldsymbol{x}_{l}^{f}(t)(\boldsymbol{x}_{l}^{f}(t))^{H}] = \sigma_{s}^{2}\boldsymbol{a}_{l}\boldsymbol{a}_{l}^{H} + \sigma_{n}^{2}\boldsymbol{I},$$

therefore the forward spatially smoothed covariance matrix is

$$\boldsymbol{R}_{F} = \frac{1}{L} \sum_{l=1}^{L} \boldsymbol{R}_{l}^{f} = \sigma_{s}^{2} \left\{ \frac{1}{L} \sum_{l=1}^{L} \boldsymbol{a}_{l} \boldsymbol{a}_{l}^{H} \right\} + \sigma_{n}^{2} \boldsymbol{I}.$$

# **Result of AOA Estimation Error** Measured result in anechoic chamber (1)



The optimal number of subarrays can be observed at L = 5 and 6.

# **Result of AOA Estimation Error** Measured result in anechoic chamber (2)



The performance of ESPRIT is improved by applying SSP.

## **Result of AOA Estimation Error** Measured result in open site



The performance of ESPRIT is improved by applying SSP.

### **Problem Formulation:** Error in Array Model

$$a_m = e^{j\phi_m}$$
 Gain and phase errors

The array output vector with error in the array response

$$\tilde{\boldsymbol{x}}(t) = \Gamma \boldsymbol{a} s(t) + \boldsymbol{n}(t)$$
  
;  $\Gamma = diag[\boldsymbol{\gamma}_1, \boldsymbol{\gamma}_2, \dots, \boldsymbol{\gamma}_M]$   
 $\tilde{\boldsymbol{x}}(t) = \tilde{\boldsymbol{a}} s(t) + \boldsymbol{n}(t)$ 

The output covaraince matrix becomes

$$\tilde{\boldsymbol{R}} = E[\tilde{\boldsymbol{x}}(t)\tilde{\boldsymbol{x}}^{H}(t)] = \sigma_{s}^{2}\tilde{\boldsymbol{a}}\tilde{\boldsymbol{a}}^{H} + \sigma_{n}^{2}\boldsymbol{I}$$

By applying FSS,  $\tilde{\boldsymbol{R}}_{F} = \frac{1}{L} \sum_{l=1}^{L} \tilde{\boldsymbol{R}}_{l} = \sigma_{s}^{2} \left\{ \frac{1}{L} \sum_{l=1}^{L} \tilde{\boldsymbol{a}}_{l} \tilde{\boldsymbol{a}}_{l}^{H} \right\} + \sigma_{n}^{2} \boldsymbol{I}.$ 

$$\tilde{a}_{m} = (1 + \tilde{g}_{m}) e^{j(\phi_{m} + \tilde{\phi}_{m})}$$
$$\tilde{a}_{m} = \gamma_{m} a_{m}; \gamma_{m} = (1 + \tilde{g}_{m}) e^{j\tilde{\phi}_{m}}$$
Assume:  $\tilde{\phi} \sim N(0, \sigma_{\phi}^{2})$ 
$$\tilde{g} \sim N(0, \sigma_{g}^{2})$$

### **Capability of SSP in reducing error (1):** Phase error case

Focusing only on the part of the array response of the above eq.;

$$B = \frac{1}{L} \sum_{l=1}^{L} B_{l} = \frac{1}{L} \sum_{l=1}^{L} \tilde{a}_{l} \tilde{a}_{l}^{H}$$

$$= \frac{1}{L} \sum_{l=1}^{L} \begin{bmatrix} 1 & e^{-j\omega} e^{-j(\tilde{\phi}_{l+1} - \tilde{\phi}_{l})} & \cdots & e^{-j(M_{0} - 1)\omega} e^{-j(\tilde{\phi}_{l+M_{0} - 1} - \tilde{\phi}_{l})} \\ e^{j\omega} e^{j(\tilde{\phi}_{l+1} - \tilde{\phi}_{l})} & 1 & \cdots & e^{-j(M_{0} - 2)\omega} e^{-j(\tilde{\phi}_{l+M_{0} - 1} - \tilde{\phi}_{l+1})} \\ \vdots & \ddots & \vdots \\ e^{j(M_{0} - 1)\omega} e^{j(\tilde{\phi}_{l+M_{0} - 1} - \tilde{\phi}_{l})} & e^{j(M_{0} - 2)\omega} e^{j(\phi_{l+M_{0} - 1} - \phi_{l+1})} & \cdots & 1 \end{bmatrix}$$

To first order of Taylor expansion:

$$\frac{1}{L}\sum_{l=1}^{L} e^{j(\tilde{\phi}_{l+1}-\tilde{\phi}_{l})} = 1 + \frac{j}{L}\sum_{l=1}^{L} (\tilde{\phi}_{l+1}-\tilde{\phi}_{l})$$
$$= 1 + \frac{j}{L} (\tilde{\phi}_{L+1}-\tilde{\phi}_{1})$$
$$\sim N(0, \frac{2\sigma^{2}}{L^{2}})$$

Phase error at p-row, q-column is distributed with

$$\boldsymbol{\zeta}_{pq} \sim N(0, 2|p-q|\frac{\sigma_{\phi}^{2}}{L^{2}})$$

for 
$$p \neq q$$

STD of the phase error reduces according to L.

# **Capability of SSP in reducing error (2)**

#### Gain error case:

Gain error at p-row and q-column is distributed with

 $\eta_{pq} \sim N(0, \frac{2\sigma_g^2}{L})$  STD of the gain error reduces according to L.

#### Tradeoff:

\* if L increases, STD of gain error decrease... But  $M_0$  also decreases. This means resolution of AOA estimation decrease. (Since resolution is proportional to no. of elements)

#### Phase error case:

Phase error at p-row and q-column is distributed with

$$\zeta_{pq} \sim N(0, 2|p-q|\frac{\sigma_{\phi}^2}{L^2})$$
 for  $p \neq q$ 

AOA estimation error caused by gain error:



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### **Simulation Condition**

Number of element (M)	16
Element Spacing (d)	Half a wavelength
AOA	6 [deg]
Signal SNR	10 dB
Number of snapshots	500
Number of trials	100

### **Simulation Result**



STD of estimated AOA by applying FSS with the number of subarrays (L) shown is smaller that that without applying FSS for both cases of gain and phase errors. 17

### **Optimal number of subarrays Measured data in anechoic chamber**



• The optimal number of subarrays can be observed and agrees with the theoretical formular the case of phase error.

• Our measured data were possibly only affected by phase errors in the array response.

### **Conclusion and Future work**

- SSP can reduce the random error in the array response occurred in the real senarios, which is verified by the improved result of the meaured-data estimation.
  - Lessening of the error of estimated AOAs.
  - Applicable in real scenarios (at least to the extent verified by our measurement system and under the field experiment conditions).
- The problem formulas according to the gain and phase error in the array response are generalized and their performances were verified by simulation.
- The optimal number of sub-arrays was also investigated to obtain the most effectiveness of SSP.
- Future works: To investigate its applicability for the multiple sources case in real systems.

*Thank you very much for your attention.* 

### What is the advantage if AOA errors reduce?

#### **Application for Mobile localization system**



Error in AOA estimation deceases (ex.  $\theta_1 - \theta_2 = 0.09$  deg.)

Error in the deteched distance deceases (  $d_1 - d_2 = 1.2 \text{ m}$ )

More precise in dectecting the required mobile terminal.